LM3444 -120VAC, 8W Isolated Flyback LED Driver

National Semiconductor
Application Note 2082
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December 7, 2010

Introduction

This demonstration board highlights the performance of a LM3444 based Flyback LED driver solution that can be used to power a single LED string consisting of 4 to 8 series connected LEDs from an 90 V_{RMS} to 135 V_{RMS}, 60 Hz input power supply. The key performance characteristics under typical operating conditions are summarized in this application note.

This is a two-layer board using the bottom and top layer for component placement. The demonstration board can be modified to adjust the LED forward current, the number of series connected LEDs that are driven and the switching frequency. Refer to the LM3444 datasheet for detailed instructions.

A bill of materials is included that describes the parts used on this demonstration board. A schematic and layout have also been included along with measured performance characteristics.

Key Features

- Line injection circuitry enables PFC values greater than 0.99
- Adjustable LED current and switching frequency
- Flicker free operation

Applications

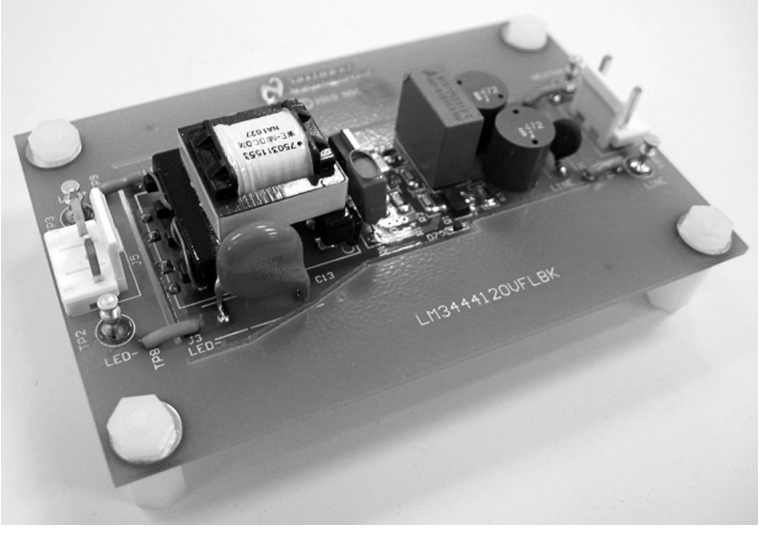
- Solid State Lighting
- Industrial and Commercial Lighting
- Residential Lighting

Performance Specifications

Based on an LED V_f = 3.57V

Symbol	Parameter	Min	Typ	Max
V_{IN}	Input voltage	90 V_{RMS}	120 V_{RMS}	135 V_{RMS}
V_{OUT}	LED string voltage	12 V	21.4 V	30 V
I_{LED}	LED string average current	-	350 mA	-
P_{OUT}	Output power	-	7.6 W	-
f_{sw}	Switching frequency	-	79 kHz	-

Demo Board



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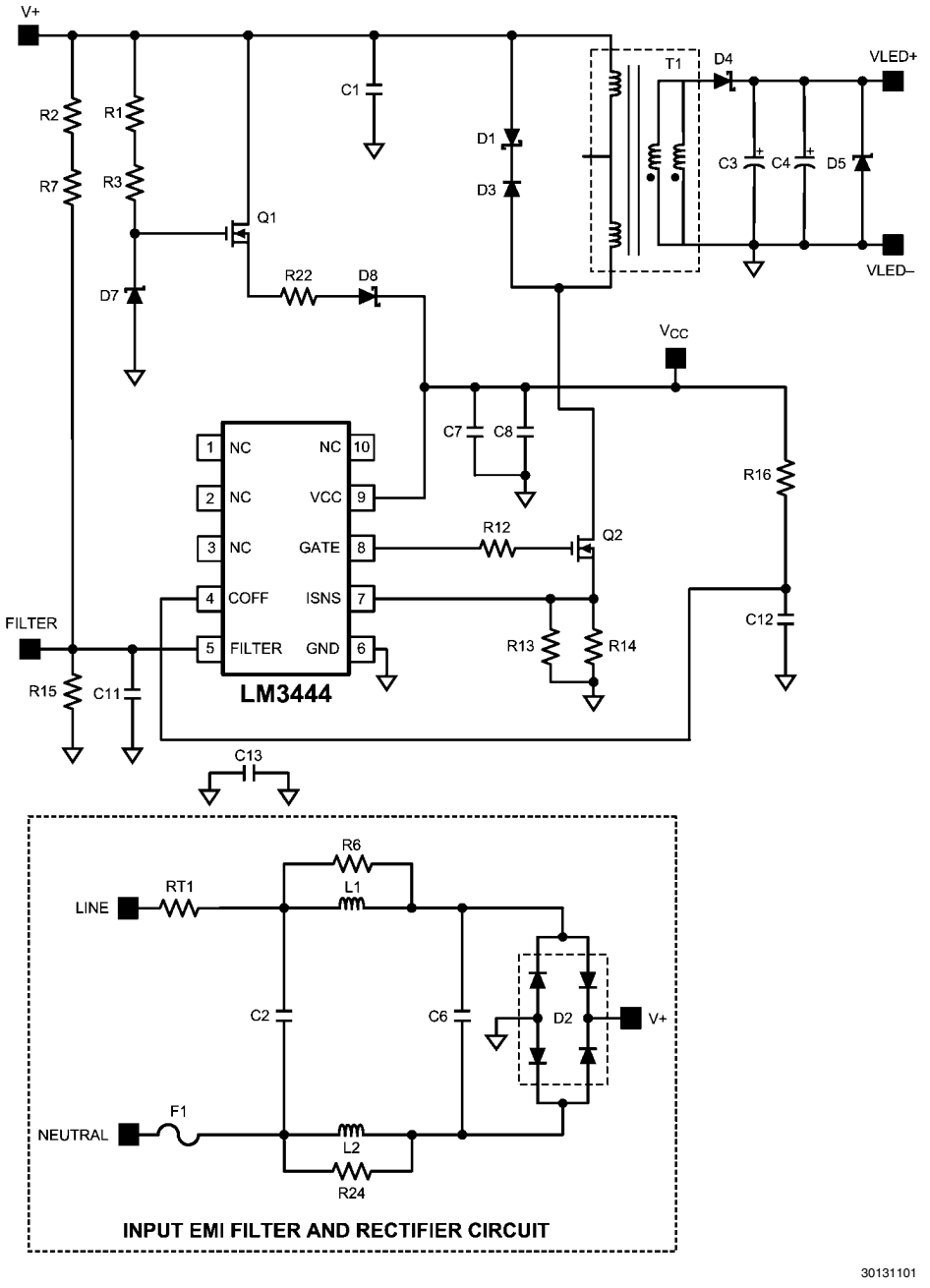
LM3444 120VAC, 8W Isolated Flyback LED Driver Demo Board Schematic

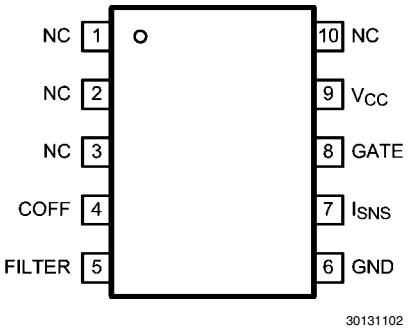


Warning: The LM3444 evaluation board has exposed high voltage components that present a shock hazard. Caution must be taken when handling the evaluation board. Avoid touching the evaluation board and removing any cables while the evaluation board is operating. Isolating the evaluation board rather than the oscilloscope is highly recommended.

Warning: The ground connection on the evaluation board is NOT referenced to earth ground. If an oscilloscope ground lead is connected to the evaluation board ground test point for analysis and AC power is applied, the fuse (F1) will fail open. The oscilloscope should be powered via an isolation transformer before an oscilloscope ground lead is connected to the evaluation board.

Warning: The LM3444 evaluation board should not be powered with an open load. For proper operation, ensure that the desired number of LEDs are connected at the output before applying power to the evaluation board.

LM3444 Device Pin-Out



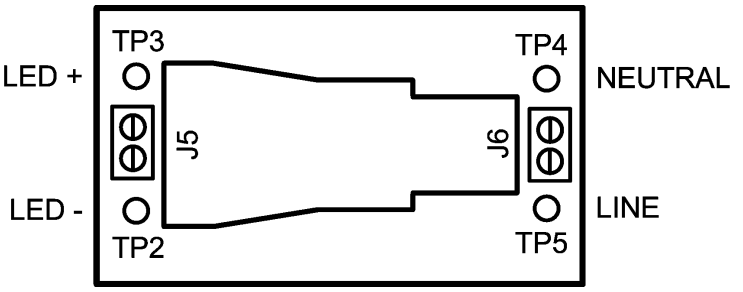
Pin Description 10 Pin MSOP

Pin #	Name	Description
1	NC	No internal connection.
2	NC	No internal connection.
3	NC	No internal connection.
4	COFF	OFF time setting pin. A user set current and capacitor connected from the output to this pin sets the constant OFF time of the switching controller.
5	FILTER	Filter input. A capacitor tied to this pin filters the error amplifier. Could also be used as an analog dimming input.
6	GND	Circuit ground connection.
7	ISNS	LED current sense pin. Connect a resistor from main switching MOSFET source, ISNS to GND to set the maximum LED current.
8	GATE	Power MOSFET driver pin. This output provides the gate drive for the power switching MOSFET of the buck controller.
9	V_{CC}	Input voltage pin. This pin provides the power for the internal control circuitry and gate driver.
10	NC	No internal connection.

Bill of Materials

Designator	Description	Manufacturer	Part Number
AA1	Printed Circuit Board	-	551600530-001A
C1	CAP .047UF 630V METAL POLYPRO	EPCOS Inc	B32559C6473K000
C2	CAP 10000PF X7R 250VAC X2 2220	Murata Electronics North America	GA355DR7GB103KY02L
C3, C4	CAP 330UF 35V ELECT PW	Nichicon	UPW1V331MPD6
C6	CAP .10UF 305VAC EMI SUPPRESSION	EPCOS	B32921C3104M
C7	CAP, CERM, 0.1µF, 16V, +/-10%, X7R, 0805	Kemet	C0805C104K4RACTU
C8	CAP CER 47UF 16V X5R 1210	MuRata	GRM32ER61C476ME15L
C11	CAP CER 2200PF 50V 10% X7R 0603	MuRata	GRM188R71H222KA01D
C12	CAP CER 330PF 50V 5% C0G 0603	MuRata	GRM1885C1H331JA01D
C13	CAP CER 2200PF 250VAC X1Y1 RAD	TDK Corporation	CD12-E2GA222MYNS
D1	DIODE TVS 150V 600W UNI 5% SMB	Littlefuse	SMAJ120A
D2	RECT BRIDGE GP 600V 0.5A MINIDIP	Diodes Inc.	RH06-T
D3	DIODE RECT GP 1A 1000V MINI-SMA	Comchip Technology	CGRM4007-G
D4	DIODE SCHOTTKY 100V 1A SMA	ST Microelectronics	STPS1H100A
D5	DIODE ZENER 30V 1.5W SMA	ON Semiconductor	1SMA5936BT3G
D7	DIODE ZENER 12V 200MW	Fairchild Semiconductor	MM5Z12V
D8	DIODE SWITCH 200V 200MW	Diode Inc	BAV20WS-7-F
F1	FUSE BRICK 1A 125V FAST 6125FA	Cooper/Bussmann	6125FA
J1, J2, J3, J4, TP8, TP9, TP10	16 GA WIRE HOLE, 18 GA WIRE HOLE	3M	923345-02-C
J5, J6	CONN HEADER .312 VERT 2POS TIN	Tyco Electronics	1-1318301-2
L1, L2	INDUCTOR 4700UH .13A RADIAL	TDK Corporation	TSL0808RA-472JR13-PF
Q1	MOSFET N-CH 600V 90MA SOT-89	Infineon Technologies	BSS225 L6327
Q2	MOSFET N-CH 600V 1.8A TO-251	Infineon Technology	SPU02N60S5
R1, R3	RES 200K OHM 1/4W 5% 1206 SMD	Vishay-Dale	CRCW1206200KJNEA
R2, R7	RES, 309k ohm, 1%, 0.25W, 1206	Vishay-Dale	CRCW1206309KFKEA
R6, R24	RES, 10.5k ohm, 1%, 0.125W, 0805	Vishay-Dale	CRCW080510K5FKEA
R12	RES 4.7 OHM 1/10W 5% 0603 SMD	Vishay-Dale	CRCW06034R70JNEA
R13	RES 10 OHM 1/8W 5% 0805 SMD	Vishay-Dale	CRCW080510R0JNEA
R14	RES 1.50 OHM 1/4W 1% 1206 SMD	Vishay-Dale	CRCW12061R50FNEA
R15	RES 3.48K OHM 1/10W 1% 0603 SMD	Vishay-Dale	CRCW06033K48FKEA
R16	RES 191K OHM 1/10W 1% 0603 SMD	Vishay-Dale	CRCW0603191KFKEA
R22	RES 40.2 OHM 1/8W 1% 0805 SMD	Vishay-Dale	CRCW080540R2FKEA
RT1	CURRENT LIMITOR INRUSH 60OHM 20%	Cantherm	MF72-060D5
T1	Transformer	Wurth Electronics	750311553 Rev. 01
TP2-TP5	Terminal, Turret, TH, Double	Keystone Electronics	1502-2
TP7	TEST POINT ICT	-	-
U1	Offline LED Driver, PowerWise	National Semiconductor	LM3444MM

Demo Board Wiring Overview



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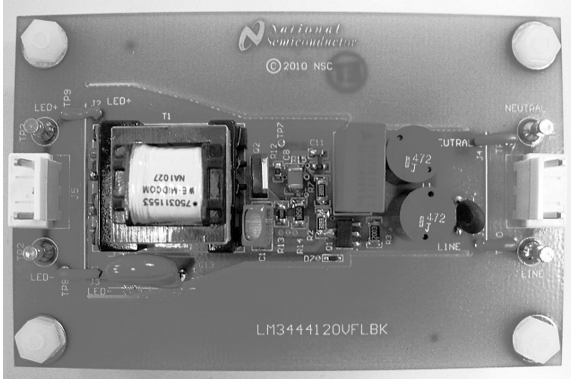
Wiring Connection Diagram

Test Point	Name	I/O	Description
TP3	LED +	Output	**LED Constant Current Supply** Supplies voltage and constant-current to anode of LED string.
TP2	LED -	Output	**LED Return Connection (not GND)** Connects to cathode of LED string. Do NOT connect to GND.
TP5	LINE	Input	**AC Line Voltage** Connects directly to AC line of a 120VAC system.
TP4	NEUTRAL	Input	**AC Neutral** Connects directly to AC neutral of a 120VAC system.

Demo Board Assembly



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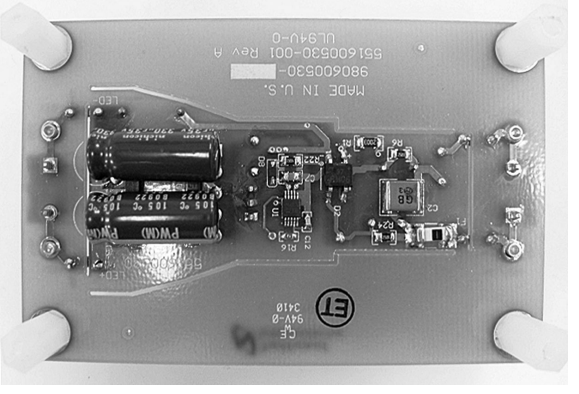
Top View



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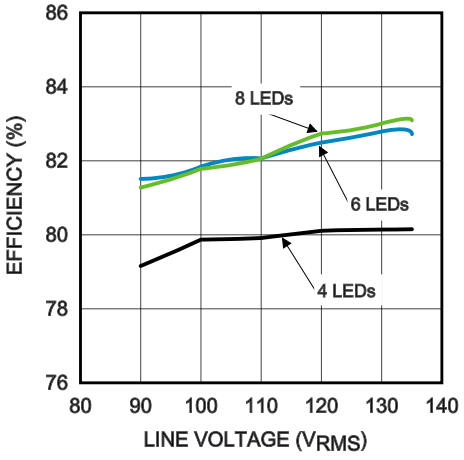
Bottom View

Typical Performance Characteristics (*Note 1*)

Efficiency vs. Line Voltage
Original Circuit



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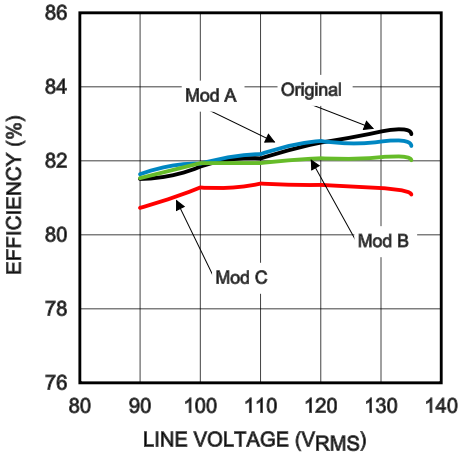
Efficiency vs. Line Voltage
Modified Circuits



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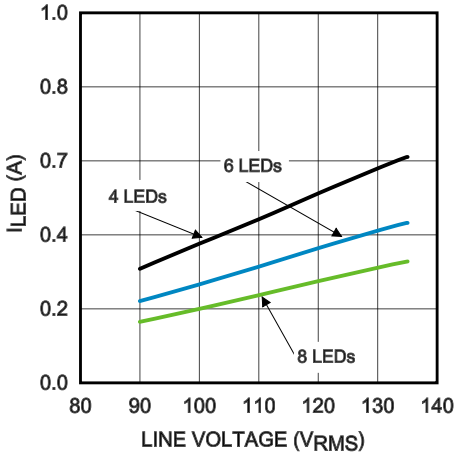
LED Current vs. Line Voltage
Original Circuit



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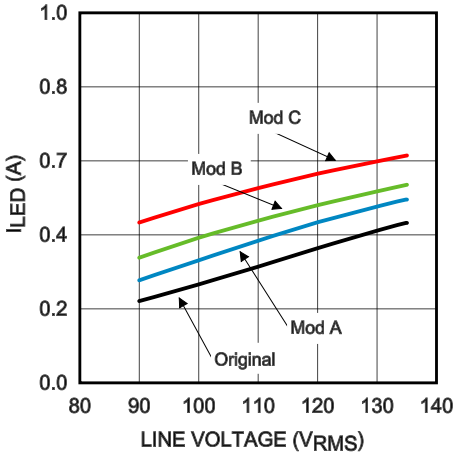
LED Current vs. Line Voltage
Modified Circuits



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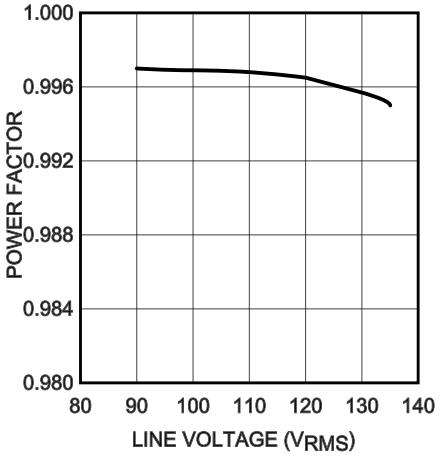
Power Factor vs. Line Voltage
Original Circuit



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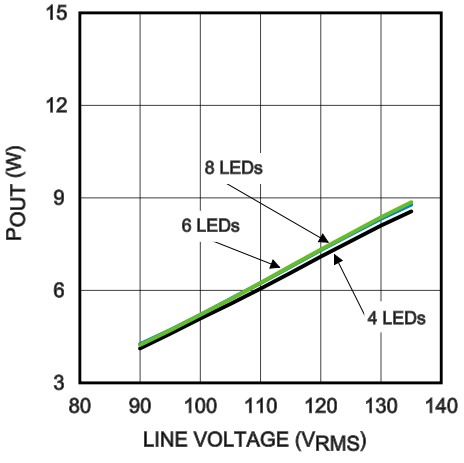
Output Power vs. Line Voltage
Original Circuit



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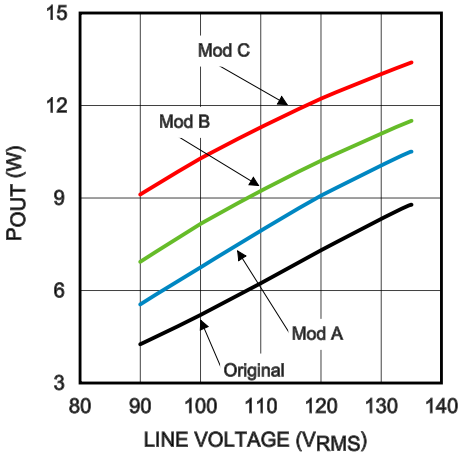
Output Power vs. Line Voltage
Modified Circuits



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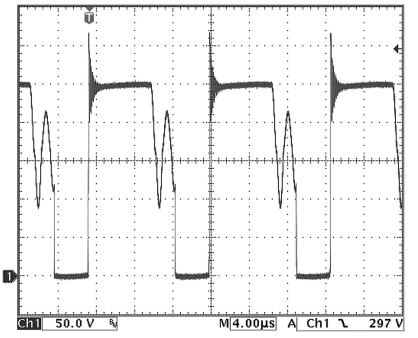
Power MOSFET Drain Voltage Waveform
(V_{IN} = 120V_{RMS}, 6 LEDs, I_{LED} = 350mA)



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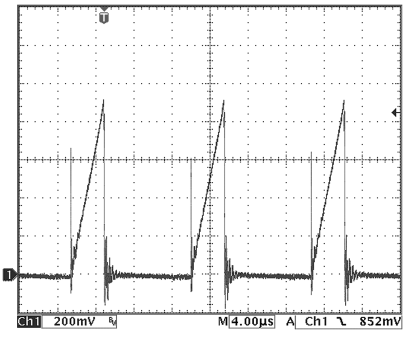
Current Sense Waveform
(V_{IN} = 120V_{RMS}, 6 LEDs, I_{LED} = 350mA)



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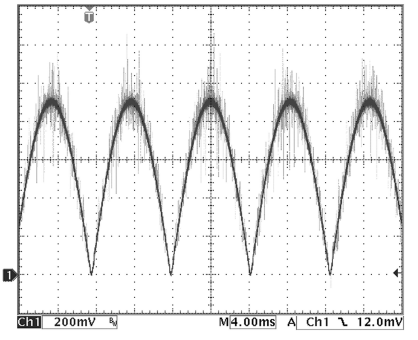
FILTER Waveform
(V_{IN} = 120V_{RMS}, 6 LEDs, I_{LED} = 350mA)



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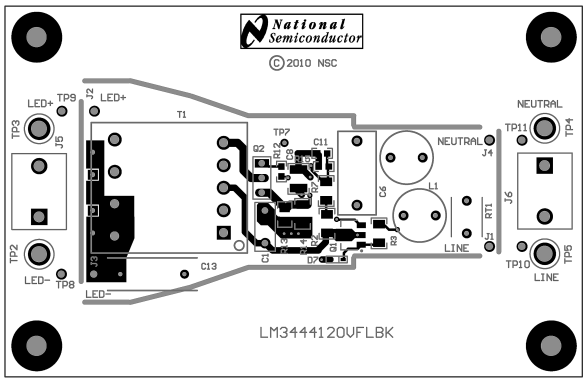
Note 1: Original Circuit: R14 = 1.50Ω; Modification A: R14 = 1.21Ω; Modification B: R14 = 1.00Ω; Modification C: R14 = 0.75Ω

PCB Layout



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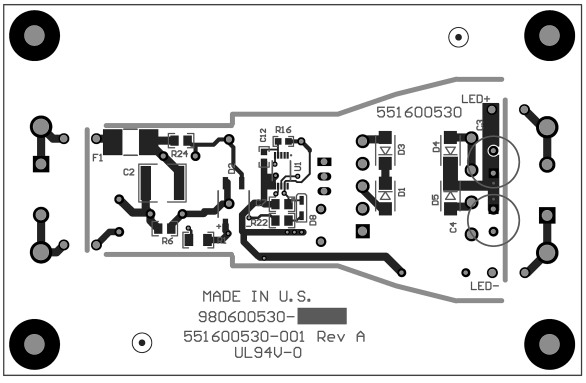
Top Layer



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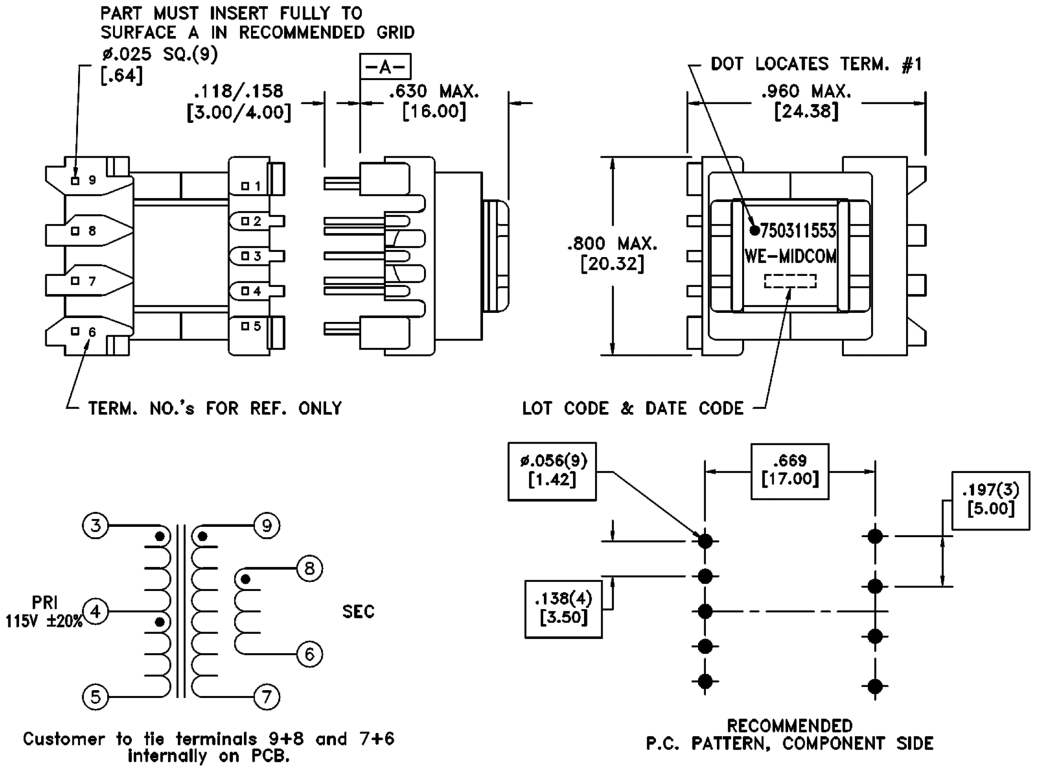
Bottom Layer

Transformer Design

Mfg: Wurth Electronics, Part #: 750311553 Rev. 01



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ELECTRICAL SPECIFICATIONS @ 25°C unless otherwise noted:

PARAMETER		TEST CONDITIONS	VALUE
D.C. RESISTANCE	3–5	@20°C	1.35 ohms max.
D.C. RESISTANCE	8–6	@20°C	0.284 ohms max.
D.C. RESISTANCE	9–7	@20°C	0.284 ohms max.
INDUCTANCE	3–5	100kHz, 100mVAC, Ls	803.5uH ±10%
INDUCTANCE	9–7	100kHz, 100mVAC, Ls	50.2uH ±10%
INDUCTANCE	8–6	100kHz, 100mVAC, Ls	50.2uH ±10%
LEAKAGE INDUCTANCE	3–5	tie(9+8+7+6), 100kHz, 100mVAC, Ls	5.50uH typ., 7.0uH max.
DIELECTRIC	3–9	tie(9+8), 4500VAC, 1 second	4500VAC, 1 minute
TURNS RATIO		(3–5):(9–7)	4:1, ±2%
TURNS RATIO		(3–5):(8–6)	4:1, ±2%

GENERAL SPECIFICATIONS:
OPERATING TEMPERATURE RANGE: −40°C to +125°C including temp rise.

Designed to comply with the following requirements as defined by IEC61558−2−17:
− Reinforced insulation for a primary circuit at a working voltage of 400VDC.

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Experimental Results

The LED driver is designed to accurately emulate an incandescent light bulb and therefore behave as an emulated resistor. The resistor value is determined based on the LED string configuration and the desired output power. The circuit then operates in open-loop, with a fixed duty cycle based on a constant on-time and constant off-time that is set by selecting appropriate circuit components.

Performance

In steady state, the LED string voltage is measured to be 21.38 V and the average LED current is measured as 357 mA.

The 120 Hz current ripple flowing through the LED string was measured to be 170 mA_{pk-pk} at full load. The magnitude of the ripple is a function of the value of energy storage capacitors connected across the output port. The ripple current can be reduced by increasing the value of energy storage capacitor or by increasing the LED string voltage.

The LED driver switching frequency is measured to be close to the specified 79 kHz. The circuit operates with a constant duty cycle of 0.28 and consumes 9.25 W of input power. The driver steady state performance for an LED string consisting of 6 series LEDs is summarized in the following table.

Measured Efficiency and Line Regulation (6 LEDs)

V_{IN} (V_{RMS})	I_{IN} (mA_{RMS})	P_{IN}(W)	V_{OUT} (V)	I_{LED} (mA)	P_{OUT} (W)	Efficiency (%)	Power Factor
90	60	5.37	20.25	216	4.38	81.6	0.9970
95	63	5.95	20.47	238	4.87	81.8	0.9969
100	66	6.57	20.67	260	5.38	81.9	0.9969
105	69	7.23	20.86	285	5.94	82.1	0.9969
110	72	7.89	21.05	309	6.50	82.3	0.9968
115	75	8.59	21.23	334	7.09	82.5	0.9967
120	77	9.25	21.38	357	7.65	82.7	0.9965
125	80	9.94	21.53	382	8.23	82.8	0.9961
130	82	10.62	21.68	406	8.80	82.9	0.9957
135	84	11.26	21.80	428	9.34	83.0	0.9950

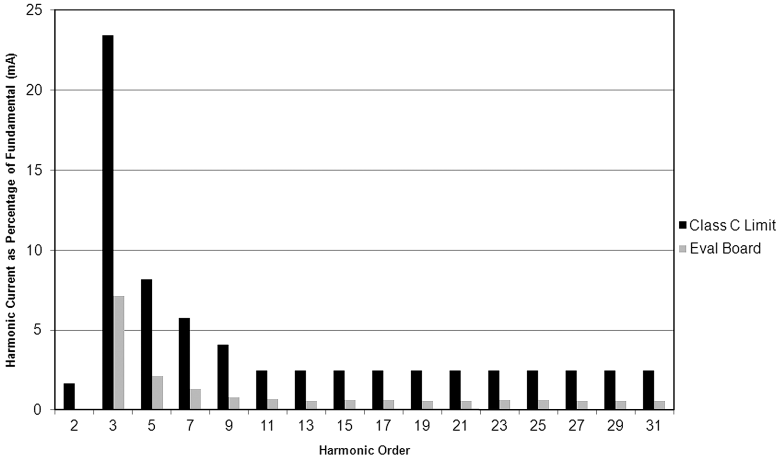
LED Current, Output Power versus Number of LEDs for Various Circuit Modifications (V_{IN} = 120 V_{AC})

# of LEDs	Original Circuit (*Note 2*)		Modification A (*Note 2*)		Modification B (*Note 2*)		Modification C (*Note 2*)	
	I_{LED} (mA)	P_{OUT} (W)	I_{LED} (mA)	P_{OUT} (W)	I_{LED} (mA)	P_{OUT} (W)	I_{LED} (mA)	P_{OUT} (W)
4	508	7.57	624	9.55	710	11.05	835	13.24
6	357	7.65	440	9.58	500	11.02	590	13.35
8	277	7.69	337	9.59	382	11.00	445	13.00

Note 2: Original Circuit: R14 = 1.50Ω; Modification A: R14 = 1.21Ω; Modification B: R14 = 1.00Ω; Modification C: R14 = 0.75Ω

Power Factor Performance

The LED driver is able to achieve close to unity power factor (P.F. ~ 0.99) which meets Energy Star requirements. This design also exhibits low current harmonics as a percentage of the fundamental current (as shown in the following figure) and therefore meets the requirements of the IEC 61000-3-2 Class-3 standard.



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Current Harmonic Performance vs. EN/IEC61000-3-2 Class C Limits

Electromagnetic Interference (EMI)

The EMI input filter of this evaluation board is configured as shown in the following circuit diagram.

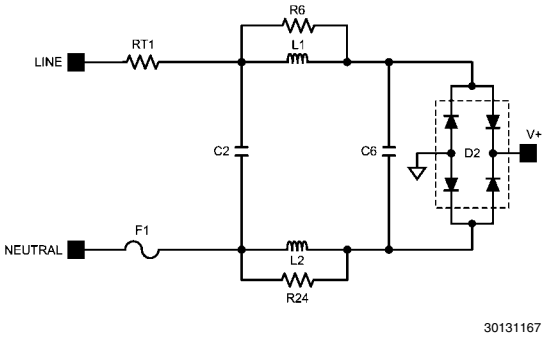

FIGURE 1. Input EMI Filter and Rectifier Circuit

In order to get a quick estimate of the EMI filter performance, only the PEAK conductive EMI scan was measured and the data was compared to the Class B conducted EMI limits published in FCC – 47, section 15.

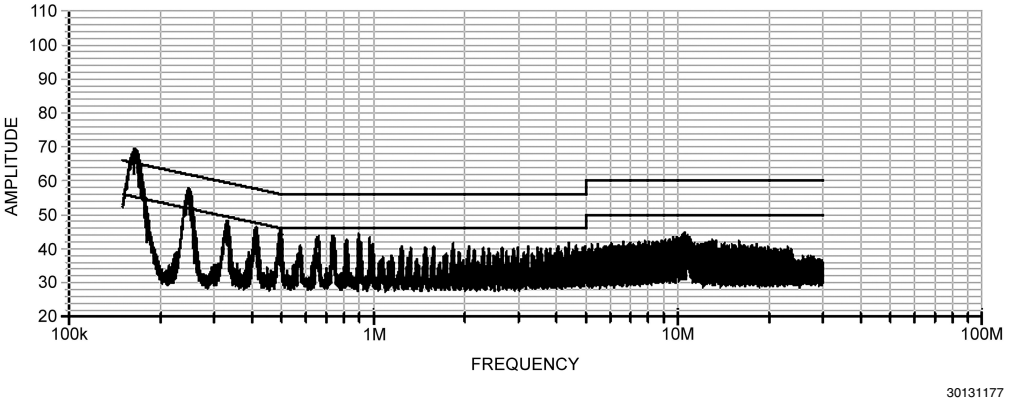

FIGURE 2. Peak Conductive EMI scan per CISPR-22, Class B Limits

If an additional 33nF of input capacitance (i.e. C6) is utilized in the input filter, the EMI conductive performance is further improved as shown in the following figure.

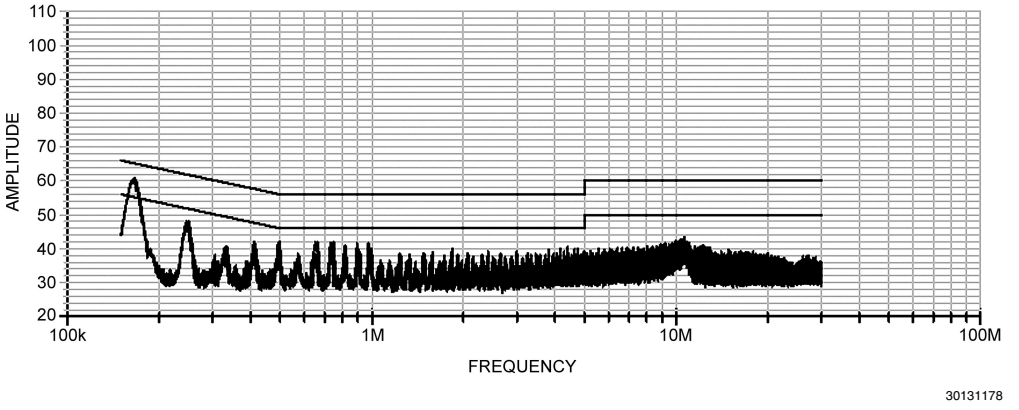

FIGURE 3. Peak Conductive EMI scan with additional 33nF of input capacitance

Thermal Analysis

The board temperature was measured using an IR camera (HIS-3000, Wahl) while running under the following conditions:

V_{IN} = 120 V_{RMS}

I_{LED} = 350 mA

# of LEDs = 6

P_{OUT} = 7.3 W

The results are shown in the following figures.

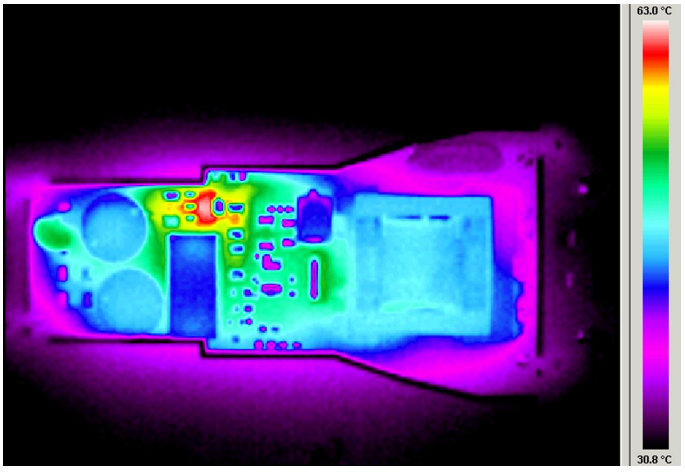

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FIGURE 4. Top Side Thermal Scan

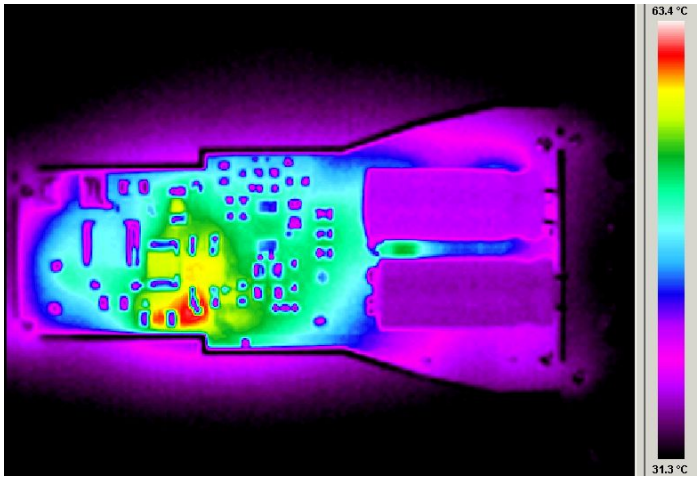

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FIGURE 5. Bottom Side Thermal Scan

Circuit Analysis and Explanations

Injecting line voltage into FILTER (achieving PFC > 0.99)

If a small portion (750mV to 1.00V) of line voltage is injected at FILTER of the LM3444, the circuit is essentially turned into a constant power flyback as shown in Figure 6.

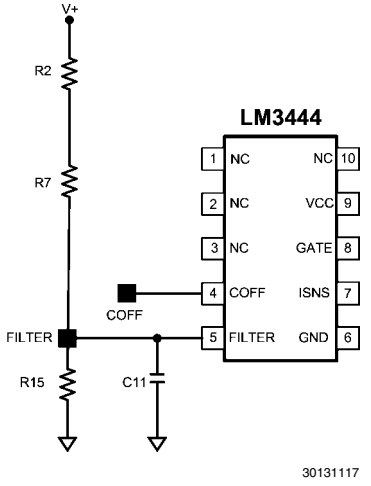

FIGURE 6. Line Voltage Injection Circuit

The LM3444 works as a constant off-time controller normally, but by injecting the 1.0V rectified AC voltage into the FILTER pin, the on-time can be made to be constant. With a DCM Flyback, Δi needs to increase as the input voltage line increases. Therefore a constant on-time (since inductor L is constant) can be obtained.

By using the line voltage injection technique, the FILTER pin has the voltage wave shape shown in Figure 7 on it. Voltage at V_{FILTER} peak should be kept below 1.25V. At 1.25V current limit is tripped. C11 is small enough not to distort the AC signal but adds a little filtering.

Although the on-time is probably never truly constant, it can be observed in Figure 8 how (by adding the rectified voltage) the on-time is adjusted.

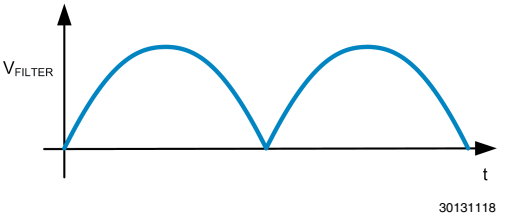

FIGURE 7. FILTER Waveform

For this evaluation board, the following resistor values are used:

R2 = R7 = 309kΩ

R15 = 3.48kΩ

Therefore the voltages observed on the FILTER pin will be as follows for listed input voltages:

For VIN = $90V_{RMS}$, V_{FILTER} = 0.71V

For VIN = $120V_{RMS}$, V_{FILTER} = 0.95V

For VIN = $135V_{RMS}$, V_{FILTER} = 1.07V

Using this technique, a power factor greater than 0.99 can be achieved without additional passive active power factor control (PFC) circuitry.

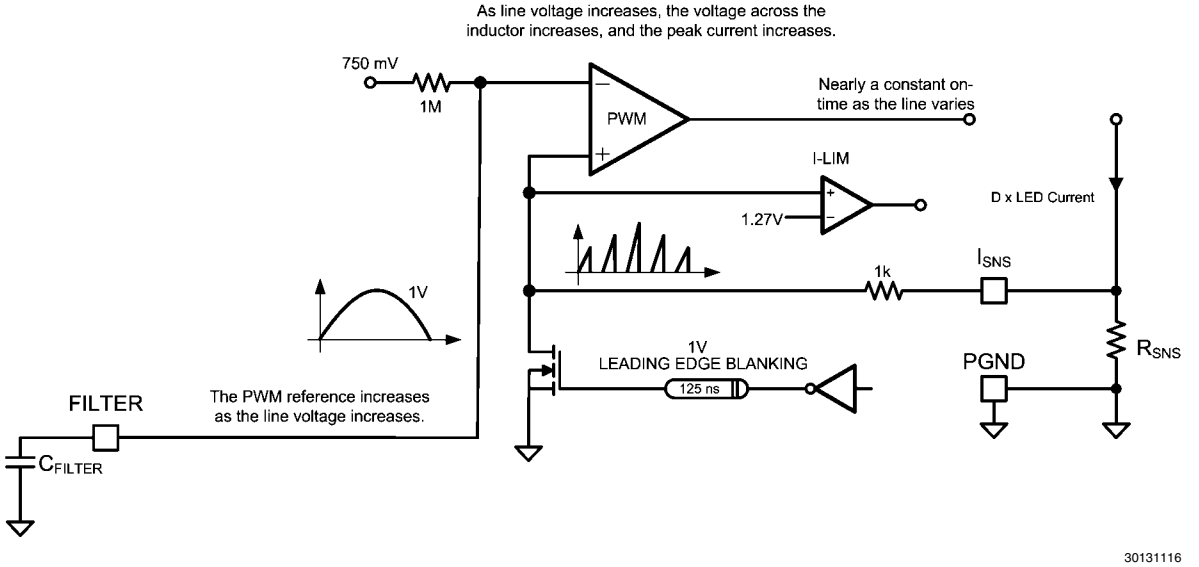

FIGURE 8. Typical Operation of FILTER Pin

Notes

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